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Analysis of a Magnetically Geared Lead Screw

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Abstract— Linear magnetic gearboxes (LMG) and magnetic lead screws (MLS) have been shown to be capable of operating at significantly higher volumetric force densities than traditional electromagnetic linear actuators (ELA). However in both such devices the linear translator must be made of magnet material and therefore if the stroke length of the translator is long the cost of the MLS and LMG will become prohibitively high. In this paper a magnetically geared lead screw (MGLS) is investigated and its performance capability is compared with the LMG and MLS. The advantage of the MGLS is that the translator does not contain magnets.

Keywords—magnetic gearbox; lead screw; linear actuator; finite element analysis

I. INTRODUCTION

Linear actuation is often achieved by utilizing either a hydraulic or mechanical gearing mechanism. By operating a hydraulic actuator at high pressure very high force densities can be sustained; for instance an excavator hydraulic piston can operate at a pressures of 38MPa [1]. Electromagnetic linear actuators have the advantage over hydraulic and mechanical mechanisms of being able to operate with higher efficiencies and are potentially more reliable [1]. However, the force density of a ELA is constrained by the current density and magnetic saturation. ELAs have been reported to have volumetric force densities in the range of 0.3kN/L with air-gap magnetic shear stress values in the range of 0.083MPa [2, 3]. Recently LMG and MLS have been proposed as a means of increasing the force densities of linear actuators. The LMG and MLS rely only on magnetic loading and therefore a higher magnetic air-gap shear stress can be sustained.

An example of a LMG is shown in Fig. 1; the LMG utilizes magnetic field heterodyning to create linear motion speed change without any physical contact [4–9]. The LMG, consists of three concentric tubular parts, an outer cylinder containing, p_o pole-pairs that can move with a translational velocity v_o an inner cylinder containing p_i pole-pairs that can translationally move at velocity v_i and a central section that contains n_t ferromagnetic rings. The ferromagnetic rings can move at velocity v_t . The ferromagnetic rings modulate the permanent magnet (PM) fields. By choosing [4]

$$n_t = p_o + p_i \quad (1)$$

the speed relationship between the translating moving parts will be [4]

$$v_t n_t = v_o p_o + v_i p_i \quad (2)$$

If $v_o=0$ the speed relationship is then given by

$$v_t = v_i (n_t / p_i) \quad (3)$$

and the gear ratio is $G_r = n_t / p_i$. Atallah and Holehouse [4–6] demonstrated that a 3.25:1 gear ratio LMG is capable of operating with a 1.7kN/L volumetric force density. By mating a stator winding with the LMG a relatively high force density magnetically geared actuator can be created [7–9]. The LMG has been investigated for use in ocean power generation applications [7, 9] as well as for vehicle suspension [8].

An example of a MLS is shown in Fig. 2. The MLS converts linear motion to rotary motion using helically shaped magnets [10–16]. The principle of operation of the MLS is analogous to a mechanical nut and screw but with a magnetic rotating “screw” and a magnetic translating “nut”. Both parts are made of helically disposed radially magnetized PMs on the inner and outer steel yokes. The relationship between translating velocity, v_t and outer angular velocity, ω_i , is given by [16].

$$v_t = k_i \omega_i \quad (4)$$

where

$$k_i = \lambda_i / (2\pi) \quad (5)$$

λ_i = inner rotor lead, which is twice the magnet pole-pitch as shown in Fig. 1. Wang calculated that the MLS could achieve

Fig. 1. An example of a linear magnetic actuator with $p_i=15$ inner pole-pairs, $p_o=6$ outer pole pairs and $n_t=21$ central ferromagnetic rings.

Fig. 2. An example of a magnetic lead screw

force densities in excess of 10kN/L [10]. Recently, Holm experimentally verified the performance of a 17kN MLS for a wave energy converter [12, 14] the calculated force density was 2.6kN/L and Berg tested a MLS for active vehicle suspension [17]. In [12, 14, 17] the rotor was driven by a secondary motor. Pakdelian [18] and Lu [16] proposed a design in which a stator could be integrating into the MLS, and so this could create a more compact design.

Both the LMG and MLS topologies require that one of the linear translating parts be made of magnet material and therefore if the linear stroke length is large then only a small portion of the magnet material will be utilized at any given time. Therefore, this will result in a low force-per-kilogram of magnet usage and consequently the design will be costly to build. The LMG force density also becomes low when the linear stroke length is increased [6]. In order to address this issue, in this paper a magnetically geared lead screw (MGLS), as shown in Fig. 3, is investigated; the important characteristic of this type of actuator is that the long-stroke translator is made of ferromagnetic material. In [19] the working concept of the MGLS was proposed. In this paper a realistically sized MGLS is designed for experimental testing and the performance capability is compared with the MLS and LMG.

The MGLS consists of three concentric tubular parts: an inner rotor with p_i , helically skewed, radially magnetized pole-pairs. An outer cylinder with p_o pole-pairs, which have been arranged in a flux-focusing structure using ring magnets and steel, and a translator which contains n_t ferromagnetic annular skewed pole pieces. As the inner rotor is helically magnetization, when it is rotated it will create a travelling field along the z -axis. This translating field will be modulated by the ferromagnetic pole pieces and therefore create additional spatial harmonics. The spatial harmonics will then interact with the stationary outer cylinder magnetic field. A constant translational force, F_z will be created only when (1) is satisfied. The rotation of the inner rotor with angular velocity, ω_i , will create a translational field velocity, v_i , given by (4). With the outer cylinder stationary the linear translator speed, v_t can be calculated from (3). For the case when $p_i=15$, $n_t=21$ the gear ratio is then $G_r=n_t/p_i=1.4$. Substituting (4) into (3) gives

$$\omega_i = v_t (n_t / (p_i k_i)) \quad (6)$$

Equation (6) relates the rotation speed of the inner rotor of the MGLS with the translational speed of the MGLS. It can be noted that (6) combines both the speed operating principals of the MLS and LMG. However, unlike the MLS and LMG the translator part is entirely made of low-cost steel. The translator

stroke length can be made very long without requiring more magnet material.

Fig. 3. Structure of the magnetically geared linear screw

Fig. 4. Cross-sectional dimensional parameters

TABLE I. FIXED GEOMETRIC AND MATERIAL VALUES

Parameter		Value	Unit
Outer cylinder (fixed) - not skewed	Pole-pairs, p_o	6	-
	Outer radius, r_{oo}	71	mm
	Pole-pitch, $2w_o$	33.6	mm
	Airgap length, l_g	0.5	mm
	Axial length, L	420	mm
Translator - annular skewed	Pole pieces, n_t	21	-
	Pole-pitch, w_t	10	mm
Inner rotor - helically skewed	Pole pairs, p_i	15	-
	Back iron, l_{ib}	11	mm
	Pole-pitch, w_i	14	mm
	Lead, λ_i	28	mm
Material	NdFeB magnet, B_r , NMX-40CH	1.25	T
	416 steel resistivity (translator)	57.0	$\mu\Omega\text{-cm}$
	1018 steel resistivity (inner, outer rotors)	15.9	$\mu\Omega\text{-cm}$

II. DESIGN ANALYSIS

In order to maximize the force density a parametric sweep has been performed. The outer radius of the outer cylinder has been set to $r_{oo} = 71\text{mm}$ so as to match that used in [12] and the axial length has been fixed at $L = 420\text{mm}$. The back iron of the inner rotor was initially made sufficiently thick so as to avoid saturation. As the geometric parameters are interrelated only three radial parameter are needed to describe the radial geometry. They are: inner radius of inner rotor, r_{ii} , outer radius of inner rotor, r_{io} , and translator outer radius, r_{to} . These parameters are shown in Fig. 4. The other geometric parameters, given in Table I, were fixed. A simple iterative technique was used to maximize the active region force density. The active region volumetric force density was computed from:

$$V_{Fd} = F_t / (\pi r_{oo}^2 L) \text{ [N/m}^3\text{]} \quad (7)$$

TABLE II. ITERATION OF RADIAL PARAMETERS

Iteration number		I	II	III	IV	V	VI	Unit
Inner rotor	Outer radius, r_{io}	53	53	51	51	48	48	mm
	Inner radius, r_{ii}	48	45.5	45.5	42	42	39.5	mm
Translator outer radius, r_{to}		58.5	58.5	57.5	57.5	54.5	54.5	mm
Translator bar thickness, l_t		5	5	6	6	6	6	mm
Outer cylinder inner radius r_{oi}		59	59	58	58	55	55	mm
Translator force, F_t		11.6	11.8	12.1	12.2	12.4	12.5	kN
Volumetric force density		1.74	1.78	1.81	1.84	1.86	1.87	kN/L
Force-per-kg magnet, F_{ko}		0.88	0.76	0.86	0.71	0.78	0.69	kN/kg

Fig. 5. Translation force as a function of inner rotor outer radius, r_{io} and translator outer radius r_{to} for iteration V. The curves for different r_{to} are shown on the figure. Fig. 6. The trade-off between maximizing volumetric force density and force-per-kg of magnet for iteration VI when $(r_{io}, r_{to}) = (48, 54.5)$ mm

The MGLS was modelling using JMAG 3-D finite element analysis (FEA) software. In iteration I r_{ii} was fixed (48 mm) and through a parametric sweep of r_{io} and r_{to} a peak force density was determined to occur at $(r_{io}, r_{to}) = (53, 58.5)$ mm. In iteration II the iteration I values were fixed and a parametric sweep for r_{ii} was performed. The results are shown in Table II. It can be seen that the peak force density is then achieved at $r_{ii} = 45.5$ mm. This iterative procedure was repeated until no significant improvement was obtained. The parametric sweep values for the last two iterations are depicted in Fig. 5 and Fig. 6. Fig. 6 shows that a trade-off exists between maximizing mass and volumetric torque density. The maximum force and force density for this design were determined to be 12.46 kN and 1.87 kN/L.

III. OPERATIONAL ANALYSIS

Using the final values given in Table II and the values in Table I the radial flux density due to the inner rotor permanent magnets (PMs) near the inner rotor and the outer cylinder have been evaluated. The results are shown in Fig. 7 and Fig. 8, the corresponding spatial harmonics, when the translator is present, is also shown. Fig. 9 and Fig. 10 shows the same plots when the PMs are only present on the outer cylinder. The modulation

effect of the translator is clearly evident. The high fundamental flux density relative to ELA designs [3] is also apparent.

Fig. 7. (a) Radial flux densities in the inner air-gap (at $r=48.1$ mm) due to magnets only on the inner rotor and (b) related harmonic values Fig. 8. (a) Radial flux densities in the outer air-gap ($r=54.9$ mm) due to magnets only on the inner rotor and (b) related harmonic spectrum values

When the inner rotor is rotated by 360° while the outer cylinder and translator are kept stationary an axial force along the z -axis is created as well as a torque. Fig. 11 shows the calculated pole slippage force and torque when using the final design parameters. It can be noted the net force on the three parts must satisfy

$$F_i + F_o + F_t = 0 \quad (8)$$

Fig. 9. (a) Radial flux densities in the outer air-gap (at $r=54.9$ mm) due to magnets only on the outer cylinder and (b) related spectrums

Fig. 10. (a) Radial flux densities in the inner air-gap (at $r=48.1$ mm) due to magnets only on the outer cylinder and (b) related spectrums

Fig. 11. (a) Pole slippage force and (b) torque on each part due to the rotation of only the inner rotor

where F_i = inner rotor force, F_t = translator force, F_o = outer cylinder force. Due to the helical structure on the inner rotor and the translator's annular skew a torque is created only on these two parts. The outer cylinder does not experience any torque since it is not skewed. The torque must satisfy

$$T_t + T_i = 0 \quad (9)$$

where T_i and T_t are the torque on the inner rotor and translator respectively. Assuming no losses the power flow relationship must then satisfy

$$F_t v_t + T_i \omega_i = 0 \quad (10)$$

and by substituting (6) into (10) and rearranging one obtains

$$T_t = -F_t k_t (p_i / n_t) \quad (11)$$

Equation (11) shows that the gear ratio reduces the torque needed to create the translational force. For example using $\omega_i = 150$ r/min and $v_t = 50$ mm/s the torque and force were calculated on the presented MGLS using 3-D FEA. The resulting torque and force values are shown in Fig. 12. It can be noted that in order to create a force of 12.46kN on the translator an applied torque of only -40Nm is required. The presented design also exhibits very low torque ripple.

IV. DISCUSSION AND CONCLUSION

This paper has numerically studied the capacities of a MGLS using comparable parameters to that used by a previously presented MLS [12]. The advantage of the proposed MGLS over prior-art designs, such as the LMG and MLS, is that the MGLS translator does not composed of magnets and therefore the cost of the translator will not significantly increase when used in long stroke applications (such as an ocean power generator).

The peak force for the presented MGLS design was determined to be 12.46kN and this corresponds to a force density of 1.87kN/L. While this is lower than for instance the 2.6kN/L and 1.95kN/L achieved by the MLS and LMG in references [12] and [6] the force-per magnet for the MGLS will not change as the stroke length increases.

Fig. 12. (a) Force and (b) torque on each part due to the rotation of the inner rotor and translation of translator at the same time.

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